CMOS Realization of VDVTA and OTA Based First Order All Pass Filter with Maximum Linearity at Low Supply Voltage of 0.85 V, Completely Electronically Tunable

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Abstract

New first order all pass filter configurations are presented in this study. The first order all pass filter configuration, based on OTAs and VDVTA, is used in the proposed all pass filter configuration. One VDVTA and one grounded capacitor are used in the first suggested design, whereas two OTAs and one grounded capacitor are used in the second suggested configuration. The quality characteristics of both types of suggested setups are independent of the tunable pole frequency range and fully electronically tunable. The reported configurations yield low active and passive sensi- tivities and also have low power consumption with very low supply voltage \pm 0.85 V with Bias Voltage \pm 0.50 V. The PSPICE simulation of the proposed VDVTA and two OTAs based first order all pass filter configurations are ve- rified using 0.18 μ m CMOS Technology Process Parameters.

Keywords

Operational Transconductance Amplifier (OTA), Voltage DifferencingTransconductance Amplifier (VDVTA), CMOS Technology

1. Introduction

Due to recent development in the fields of microelectronics, analog signal processing as well as digital signal processing and all the fields of communication, size of the transistors and power supplies are reduced. Last few decades, current mode active building blocks have been evolved in the realization of ac-

tive filters and oscillators. Current mode building blocks have higher bandwidth, higher dynamic range and higher slew rate with low power consumption. First order all pass filter and higher order filters are widely used in the fields of analog signal processing, measurement, instrumentation, voice or audio frequency range. The magnitude characteristics play vital role due to insensitivity of ear and Biomedical applications. Recently, most widely active building blocks are used in the designing oscillators, grounded Inductor and active filter applications. These active building blocks namely operational Amplifier, Current-mode current, gain first-order all pass filters employing CFTAs [1], A voltage-mode first order all pass filter based on VDTA [2], Voltage-mode all-pass filters in- cluding minimum component count circuits [3], Voltage-mode cascadableall-pass section using single active element [4], Single VDVTA Based Vol- tage-Mode Biquad Filter [5], Single MO- CCCCTA-Based Electronically Tuna- ble Current/Trans-Impedance-Mode Biquad Universal Filter [6], Electronically Tunable Low Voltage Mixed-Mode Universal Biquad Filter [7], Current-tunable current-mode all-pass section using DDCC [8], Electronically tunable first-order all pass section using OTAs [9], Current-mode multi phase sinusoidal oscillator using CDTA-based all pass sections [10], New resistorless and electronically tunable realization of dual-output

VM all-pass filter using VDIBA [11], Cascad-able Current-mode first order all-pass filter based on minimal components [12], Voltage-mode all-pass filters using universal voltage conveyor and MOSFET-based electronic Resistors [13], Voltage mode cascadable all-pass sections using single active element and grounded passive components [14], Electronically tunable first-order all-pass circuit employing DVCC [15], Novel voltage-mode all-pass filter based on using DVCCs [16], A Resistorless realization of the first-order all-pass filter [17], High input impedance voltage-mode first-order all-pass sections [18], Unity/variable-gain voltage-mode/current-mode first-order all-pass filters using single DXCCII [19], First-order voltage-mode all-pass filter employing one active element and one grounded capacitor [20], Component reduced all-pass filter with a grounded capacitor and high-impedance input [21], Realization of Grounded Inductor Based Band Pass Filter Design to Achieve Optimum Linear-ity with Bandwidth using Single VDVTA [22], Multi output filter and four phase sinusoidal oscillator using CMOS DX-MOCCII [23], FDCCII based Electroni-cally Tunable Voltage Mode Biquad Filter [24], CDBA Based Inverse Filter [25].

2. Description of the Proposed Active Building BlocksVDVTA and OTA for First Order All Pass Filter

The symbolical representation of the VDVTA as an active element is shown in **Figure 1** contains three input terminals namely P, N, V and Z, X^+ , X^- are outputterminals. All input terminals and output terminals exhibit high impedance val-ues. The symbolic representation of OTA is shown in **Figure 2(a)** and **Figure 2(b)** represents Equivalent Circuit of OTA respectively. The symbolical repre-sentation of the VDVTA as an active element is shown in **Figure 1** contains three input terminals namely P, N, V and Z, X^+ , X^- are output terminals. All

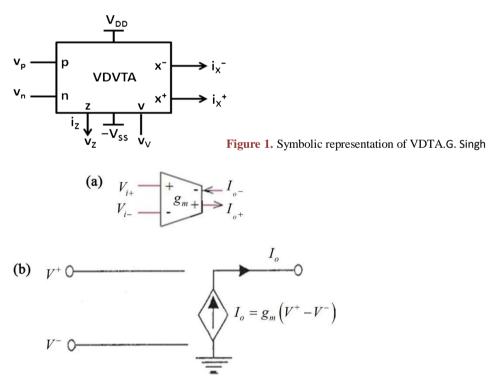


Figure 2. (a) Symbolic Notation of OTA; (b) Equivalent Circuit of OTA.

input terminals and output terminals exhibit high impedance values.

The characteristics equations of the proposed VDVTA as active building blocks for realization of first order all pass filter can be reported as Equation (1.0):

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 $I_{x+} = g_m \left(V_z - V_v \right) I_{x-} = -g_m \left(V_z - V_v \right) I_z = g_m \left(V_p - V_n \right)$ The currents $x + 2 \cdot I_{x-}$ and I_z are characterized by the Equations (1.1).

(1.2) and (1.3). The operational transconductance amplifier is the basic building block for realizing all pass filter and I_o is the output current of proposed OTA as an active building block. All terminals of OTA exhibit high impedance values. For an ideal OTA the transconductance g_m and the output current I_o is characte-rized by Equation (2.0) $_o = g_m (V_+ - V_-)$ (2.0)The transfer functions of transadmittance mode all pass filter are employed single VDVTA with one grounded capacitor is shown in **Figure 3** and two OTAs with one grounded capacitor is shown in **Figure 4**.

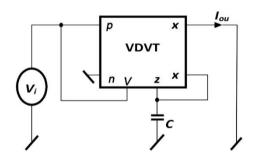


Figure 3. Proposed VDVTA based all pass filter.

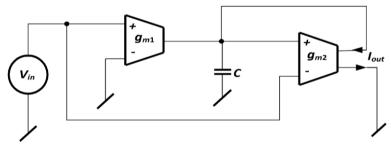


Figure 4. Proposed OTA based all pass filter.

The transfer functions of transadmittance mode all pass filter is given in Equ-ation (2.1)

$$\frac{\left|\frac{I_{out}}{V_{in}}\right|}{\left|\frac{I_{out}}{V_{in}}\right|} \qquad g \qquad \qquad = \frac{\left(\begin{array}{c}g_{m1} - SC\\(2.1)g\end{array}\right)}{\left(\begin{array}{c}APF - m2SC\end{array}\right)}$$

The proposed transimpedance mode all pass filter using two OTAs.

OTAs contains high impedance input terminals and the implementation of the proposed transimpedance mode all pass filter using single VDVTA and onegrounded capacitor respetively.

Transfer function of proposed OTAs based all pass filter is characterized bythe Equation (2.2):

$$\left| \frac{V_{out}}{V_{in}} \right| = \frac{G_{ml}}{+SC} + 2g$$
(2.2)

APF m^2 1 0

The phase margin and gain of the proposed all pass filters using two VDVTA and OTAs can be described in Equation (2.3):

$$V(s)$$

$$V_{in}(s)$$

$$(\varpi) = 180^{\circ} - 2 \tan^{-1} (\varpi RC)(2.3)$$

3. CMOS Simulation Result

The CMOS Simulation result shows that the workability and functionality of the proposed VDVTA and OTAs based first order all pass filters have been verified using PSPICE. The aspect ratios of MOS transistor for OTA and VDVTA are reported in **Table 1**. The CMOS realization of the proposed VDVTA and OTAs based first order all pass filters are shown in **Figure 5** and **Figure 6** respectively. The proposed first order filter configurations are biased with VDD = -VSS =

Table 1. Aspect Ratio of MOS Transistors for OTA and VDVTA.

OTA

MOS Transistor	Width		VDVTA			
		Length	MOS Transistor	Width (µm)	Length	
	(μ m)	(μ m)			(µm)	
M1, M3, M5, M7	8.2 4	0.3 6	M1, M3, M5, M7, M9, M11	3.6	0.6	
M2, M4, M6	7.2	0.3 6	M2, M4, M6, M8, M10, M12	8.24	0.6	
M8, M9	14. 4	0.3 6	M13, M15, M17	7.2	0.6	
M10, M11	3.6	0.3	M16, M18	3.06	0.6	

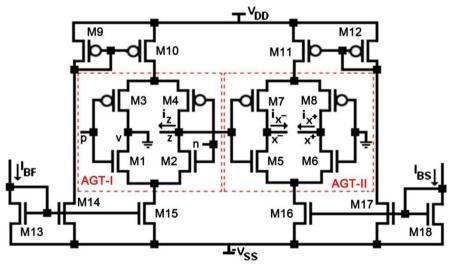


Figure 5. CMOS Realization of VDVTA Based All Pass Filter.

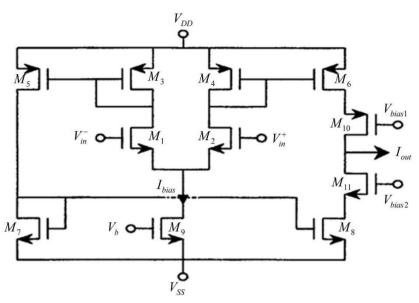


Figure 6. CMOS Realization of OTA for the proposed First order all pass filter.

 ± 0.5 V with transconductances at different bias currents are $g_{m1} = g_{m2} = 68.96$ μ A/V - 103.7 μ A/V for $I_{Bias} = 10$ μ A, $g_{m1} = g_{m2} = 86.24$ μ A/V - 104.65 μ A/V for $I_{Bias} = 20$ μ A, $I_{Bias} = 30$ μ A, $g_{m1} = g_{m2} = 789$ μ A/V for $I_{Bias} = 50$ μ A - 70 μ A, $g_{m1} = g_{m2} = 789$ μ A/V for $I_{Bias} = 50$ μ A - $I_{Bias} = 10$

 $g_{m2} = 68.16 - 734 \,\mu\text{A/V}$ for $I_{Bias} = 100 \,\mu\text{A} - 150 \,\mu\text{A}$, $g_{m1} = g_{m2} = 86.16 - 734 \,\mu\text{A/V}$ $I_{Bias} = 170 \,\mu\text{A} - 300 \,\mu\text{A}$. The simulated gain magnitude response 24 dB - 35.13 dB of OTAs based first order all pass filter is in **Figure 7** and the transient response of input and output for proposed first order all pass filters configuration is presented for 0 - 60 mV input and output signal as shown in **Figure 8**. The **Table 2** presents the comparison of the proposed first order all pass filter configurations. The percentage of total harmonic distortion for the input of proposed first order all pass filters configuration is observed 1.21% - 2.83% with minimum power dissipation 1.6 mW - 4.76 mW as shown in **Figure 9** and the simulated phase margin of VDVTA based first order all pass filter is approximately from 0 to 180 and 0 to 360 deg or vice versa as shown in **Figure 10**. The pole frequency varies with 90 deg at 1.53 KHz - 10.498 MHz The simulated frequency response of proposed all pass filter configuration is presented in **Figure 11** which yields open loop gain of is 5.045 - 6.020 dB and typical values 30 dB - 71 dB with 4.10

 $\sqrt{\text{Hz}}$ at the input and 3.8 $\text{nV}/\sqrt{\text{Hz}}$ noise spectral density at the cut off

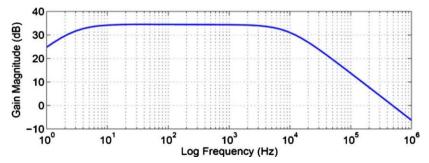


Figure 7. Simulated gain magnitude response of proposed OTA based all pass filter.



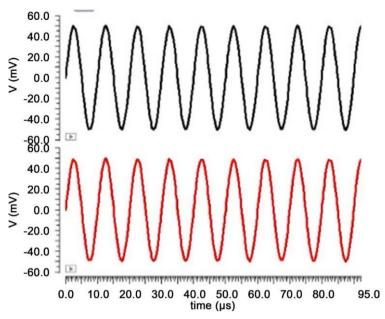


Figure 8. Transient response of the input and output for proposed VDVTA and OTAbased all pass filter.

Table 2. Comparison of the proposed All Pass filter Configuration With various All Pass filter configuration.

	Ref No				No of A	ctiveComponents
No Passive Componentsfloating or Groun	nded:		R - C			Condition
Grounded Capacitor	1	CFTA	1 Resistor, 1 Capacitor	Yes	Yes	No
T1	2	VDTA	1 Capacitor	No	Yes	No
Electronic tunability	8	DDCC	3 Resistor, 1 Capacitor	Yes	No	No
ComponentMatching	9	OTA	1 Capacitor	Yes	Yes	No
	10	CDTA	1 Capacitor	No	Yes	No
	11	VDIBA	1 Capacitor	No	Yes	No
	24	FDCCII	2 Resistor, 1 Capacitor	Yes	No	No
	25	CDBA	3 Resistor, 1 Capacitor	No	No	Yes
	Proposed Work	VDVTA, OTA	1 Capacitor	Yes	Yes	Yes

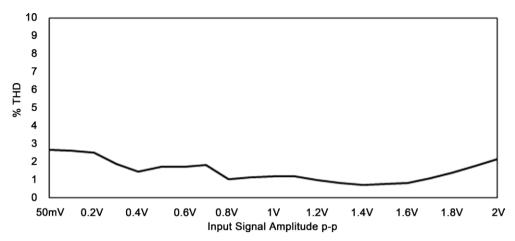


Figure 9. % THD at the input signal of all pass filter.

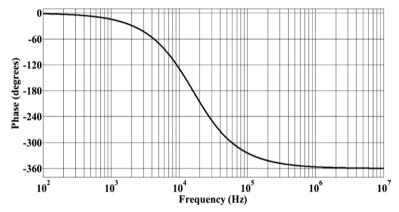


Figure 10. Simulated phase margin of proposed all pass filter configurations.

frequency 1.53 KHz - 10.498 MHz and Theoretical (F_{in}) approximately 1.499KHz with power supply ± 0.85 V and Bias voltage ± 0.5 V.

The device model parameters are taken from TSMC 0.18 µm have been simu-

lated in PSPICE with high linearity at ± 0.5 V Supply Voltage at different bias current. The simulated parameters of the proposed VDVTA and OTAs based first order all pass filters are reported in **Table 3**. are taken from TMSC 0.18 μ m CMOS technology parameters at ± 0.5 V supply voltages all gm equal to 639.7 μ A/V or 734 μ A/V are used for the proposed VDVTA and OTAs based first or- der all pass filters yield high linearity.

4. Performance Evaluation

The performance of the proposed VDVTA and OTAs based first order all pass

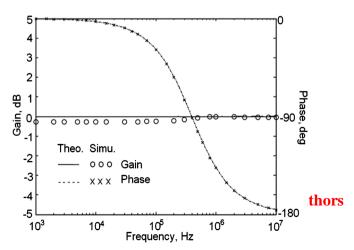


Figure 11. Simulated frequency response with phase of proposed all pass filter.

Table 3. Simulated Parameters of Proposed VDVTA and OTA based All Pass Filter at Different Bias Currents: $I_{Bias} = 10 \,\mu\text{A} - 300$ μΑ.

S. No	Specifications	Simulated	
1	CMOS Technology	0.18 μm	
2	Transconductances ($\mu A/V$)	68.16 μA/V - 789 μA/V at 10 μA - 300	μΑ
3	Bias current (µA)	10 μΑ - 300 μΑ	
4	% Total Harmonic Distortion	1.21% - 2.83%	
5	Power dissipation (mW)	1.6 mW - 4.76 mW	
6	Maximum Input noise (nV)	33 nV- 42 nV	
7	Maximum output noise (nV)	24 nV	
8	Maximum Supply Voltage (V)	±0.85 V	
9	Bias Voltage (V)	±0.50 V	
10	Phase Margin	23-45	
11	Input output Voltage Swing (mV)	0 - 60 mV	
	5.045 - 6.020 dB with typical val	ues 30 dB - 71	
		dB	
	1.53 KHz and 10.498 MHz		
	at the input and 3.8 nV	$/\sqrt{\text{Hz}}$	√Hz

filters are taken from TMSC 0.18µm CMOS technology parameters. The both types of configurations yield low noise and high stability at the desired range of pole frequencies. Transconductance gain is directly proportional to the bias cur-rent. The simulated results of all pass filters by OTA observed agree with the si-mulated results of VDVTA based all pass filter.

5. Conclusions

Open loop Gain (dB)

Noise Spectral Density4.10 nV

Frequency (KHz)

The proposed OTAs and VDVTA-based all pass filter's viability and functioning are confirmed using CMOS simulation in the SPICE 0.18 um CMOS Technology. First order all pass filters are frequently used in analogue signal processing to change the phase of the input signal. The reported OTAs and VDVTA based all pass filter yields some advanta-geous features:

- 1) Fully integrated with monolithic ICs.
 - Electronic tunability with different pole frequencies at different bias cur-2) rents.
 - The proposed VDVTA based all pass filter configuration exhibits low input impedance and high output impedance.
 - 4) The proposed VDVTA based all pass filter configuration exhibits good sensitivity performance.
 - The reported work represents wideband bulk-driven OTA with tunable transconductance was designed.
- 6) The natural frequency (ω_0) of the reported OTAs and VDVTA based all pass filter can be tuned with g_{m2} or gm_2 and with capacitor C2 at constant band-width.
 - Bandwidth can be controlled with the help of transconductance (g_{m1}) and quality factor (Q) can be controlled independent of natural frequency (ω_0) with transconductance (g_{m1}).
 - For the first configuration using single VDVTA, we can change cut off fre-8)

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quency or pole frequency by changing g_m .

- 9) Whereas in second configuration OTAs based all pass filter, the cut off frequency or pole frequency can be changed by tuning g_{m1} or g_{m2} .
- 10) The integration of proposed OTA and VDVTA based first order all pass filter is open area in future research using modern mixed analog signal processing based integrated circuits for achieving high linearity at low voltage.

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